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Image reconstruction using spectroscopic and hyperspectral information for compressive terahertz imaging

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Terahertz (THz) time-domain imaging is an emerging modality and has attracted a lot of interest. However, existing THz imaging systems often require a long scan time and sophisticated system design. Recently, a new design incorporating compressed sensing (CS) leads to a lower detector cost and shorter scan time, in exchange for computation on a set of experimental THz data captured in a single-pixel THz system. © 2010 Optical Society of America

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1. INTRODUCTION

In recent years, advances in terahertz (THz) science have attracted increasing attention in time-domain THz imaging for a very diverse array of applications. These various potential applications range from biology and medical sciences [1] and non-destructive evaluation [2] to quality control processes [3] and homeland security [4]. In contrast to the conventional optical modality that is based on relative intensity measurements only, one important advantage of THz time-domain imaging systems is that the transient electric field itself can be measured, from which we can determine the amplitude and phase of each spectral component that makes up the entire THz pulse. Thus, THz imaging systems provide opportunities for spectroscopic studies, which are not possible with many competing modalities. Unfortunately, most existing THz imaging systems suffer from slow acquisition because of their raster-scanning nature [2,5,6]. For example, using one of the fastest raster-scanning time-domain THz imaging system to date [6], we still need 6 min to scan a 100 mm² area at 0.25 mm resolution (equivalently, a 400 × 400 pixel image). Although recent efforts by using more sophisticated imaging technologies, such as interferometric and tomographic approaches, have shown preliminary successes, raster-scanning is still a latent factor that limits the acquisition rate of such systems. The high complexity and hardware requirements are also major limitations to their practical applications.

To meet the requirements of practical, time-crucial applications, Chan et al. [7,8] propose two fast time-domain THz imaging schemes by applying the compressed sensing (CS) theory [9,10]. In [7], the imaging process is accelerated by randomly sampling only a subset of the Fourier amplitude measurements. However, signal acquisition in this system still requires the THz receiver to raster scan the focal plane. To replace this mechanical raster-scanning, they then present a proof-of-concept single-pixel THz imaging setup in [8]. The system schematic is illustrated in Fig. 1. Instead of collecting the pixels/voxels, a single detection element is used to sample the concentrated THz beams, which are spatially modulated with a set of random patterns. According to the CS theory, much fewer samples than the total number of image pixels are needed to fully reconstruct an image, which thus implement fast compressed imaging. In exchange for high-speed signal acquisition, the challenge of implementing such a system comes in developing efficient sparsifying representations and the corresponding algorithms for signal recovery from the linear measurements. The CS theory suggests the conditions that need to be fulfilled to allow nearly perfect reconstruction with a much smaller number of measurements. However, in reality, these conditions are hard to satisfy fully. In addition, the signal obtained from time-domain THz imaging systems is the transient electric field. That is, the received signal is complex [8,11]. Several CS reconstruction algorithms have been proven effective in dealing with real-valued signals, but for complex-valued signals, challenges still exist. The authors in [12,13] extend the considerations to the complex domain. However, aside from the $\ell_1$ norm regularizer on the complex data, they did not exploit the other a priori information.
practical THz data. Last, based on the experimental re-
onstrate in Section 3 the output of our algorithm on the
hyperspectral image estimation and devise a hyperspec-
tral sparse reconstruction method with multiscale
valued data. In Subsection 2.D, we combine the single-
spectral image estimation approach for the complex-
2.C we introduce our partition-based multiscale hyper-
more technical details are provided. Then, in Subsection
the amplitude intensity, any other
amplitude of \( \mathbf{x} \). So aside from sparsity of the am-
its not commonly found in other ones: the THz pulse con-
tains information at multiple frequencies, and therefore
hypserspectral information [14,15]. We thus consider
further enhancing the reconstruction image quality by
making use of such hyperspectral information. Prelimi-
ary work along this direction has been reported in [16].

This paper is organized as follows. Following a brief re-
view in Subsection 2.A concerning the CS theory and the
reconstruction techniques commonly used, we propose a
sparse reconstruction method for signal recovery at a spe-
cific frequency in Subsection 2.B. The work presented
here is an extension of our previous work in [17], and
more technical details are provided. Then, in Subsection
2.C we introduce our partition-based multiscale hyper-
spectral image estimation approach for the complex-
valued data. In Subsection 2.D, we combine the single-
bond sparse reconstruction method with multiscale
hyperspectral image estimation and devise a hyperspec-
tral image reconstruction algorithm for the compressive
THz imaging system. For illustration purposes, we dem-
strate in Section 3 the output of our algorithm on the
practical THz data. Last, based on the experimental re-
results, we draw our conclusions in Section 4.

2. METHODS
A. Compressed Sensing Background
Compressed sensing is concerned with encoding a sparse
signal using a relatively small number of linear measure-
ments and ensures accurate reconstruction with a very
high probability when a sufficient condition called the re-
stricted isometry property (RIP) is satisfied [9,10]. Math-
ematically, given an \( M \times N \) measurement matrix \( \Psi \) with
\( M < N \) and a vector \( \mathbf{b} \) that denotes the linear measure-
ments of an \( N \)-dimensional sparse signal \( \mathbf{x} \), say \( \mathbf{b} = \Psi \mathbf{x} \),
the optimal estimate of the underlying sparse signal can be solved by

\[
\text{minimize } \| \mathbf{x} \|_0 \quad \text{subject to } \Psi \mathbf{x} = \mathbf{b},
\]

with \( \| \mathbf{x} \|_0 \) the number of nonzero entries in \( \mathbf{x} \), also called the \( \ell_0 \) norm. That is, we can estimate the underlying sparse/compressible signals (e.g., THz signals) by searching the sparsest solution satisfying \( \Psi \mathbf{x} = \mathbf{b} \). But this is computationally intractable and highly sensitive to noise [18]. Consequently, an easier-to-solve linear program is used for searching the sparsest solution of \( \Psi \mathbf{x} = \mathbf{b} \); that is,

\[
\text{minimize } \| \mathbf{x} \|_1 \quad \text{subject to } \Psi \mathbf{x} = \mathbf{b},
\]

where \( \| \mathbf{x} \|_1 = \sum_{i=1}^{N} |x_i| \) denotes the \( \ell_1 \) norm of a vector \( \mathbf{x} \). In
other words, one minimizes the \( \ell_1 \) norm of the signal instead of the sparsity itself.

To solve such an inverse problem, several methods and
code packages have been developed, such as the SPGL1
algorithm [19,20], gradient projection for sparse recon-
struction [21], \( \ell_1 \)-magic [22], etc. Since the signal \( \mathbf{x} \) con-
sidered in the conventional CS theory is real, most of the
algorithms focus only on processing real-valued signals.
Although some researchers have tried to extend their
methods to the complex domain [19], the reconstruction
quality in practical applications is often not satisfactory.
For instance, one such attempt can be observed from Fig.
4 in [8]. If we consider the optimization problem in Eq. (2)
carefully, we may note that for the complex case (i.e., \( \mathbf{x} \in \mathbb{C}^N \)), the term \( \| \mathbf{x} \|_1 \) actually imposes constraints only on
the amplitude of \( \mathbf{x} \). So aside from sparsity of the am-
itude intensity, any other \textit{a priori} information (e.g., the
phase) has not been exploited. The pulsed THz imaging
systems are well-known for providing spectroscopic phase
information, and this leads us to further considerations
on the use of the phase. On the other hand, in the time-
domain setups, such as the THz imaging systems in [7,8],
some underlying relations have not been used for further
improvement in the image reconstruction. This includes
the similarities between spatial intensity distributions of
the amplitude and phase and the correlation across the
hyperspectral bands. In the following sections, we demon-
strate the capabilities of this prior knowledge in the re-
covered image quality improvement.

B. Single-Band Compressed Sensing Reconstruction
The single-pixel pulsed THz imaging system in Fig. 1
implements a CS process. The basic principle behind such
a CS imaging system can be described as, at a particular
frequency \( f_k \),

\[
\mathbf{b}(f_k) = \Psi \mathbf{x}(f_k),
\]

where \( \mathbf{b}(f_k) \in \mathbb{C}^M \) is a column vector of measurements and
\( \mathbf{x}(f_k) \) represents the underlying \( N \times N \) complex-valued sig-
nal ordered in an \( N^2 \times 1 \) vector, spatially modulated by a
set of random patterns that form the measurement ma-
trix \( \Psi \in \mathbb{R}^{M \times N^2} \). As mentioned above, the CS theory
provides us with the mathematical basis for the accurate
recovery of the original signals with only a few measure-
ments (i.e., \( M \ll N^2 \)). For simplicity, one could directly
adapt the optimization program in the form of Eq. (2) to
our case, just recognizing that the true and observed data
are both complex. Since seeking the sparsest exact solu-
tion may be useless because of the additive noise, an ap-
proximate reconstruction is often preferred. It then makes sense to replace Eq. (2) by
\[
\text{minimize } |\mathbf{x}(f_i)|_1 \quad \text{subject to } \| \Psi \mathbf{x}(f_i) - \mathbf{b}(f_i) \|_2 \leq \epsilon,
\]
with \( \epsilon \) the tolerance to be defined. Unfortunately, such a simple extension to the complex domain can hardly produce a satisfactory estimation of the original scene. An example of this can be found in Fig. 4 in [8], and also in our experimental results in Section 3.

In our recent work [17], we propose an effective method that can significantly improve the reconstruction quality in combination with the phase information carried in the received THz pulses. The basic consideration relies on the fact that the phase should be smooth and not vary rapidly [23, 24]. Hence, to interpret the smoothness constraint on the phase, we apply the similarities between the \( i \)th pixel and the spatial neighborhood \( \Omega_i \) of a certain size centered at pixel \( i \). The size of \( \Omega_i \) controls the degree of spatial smoothness. Meanwhile, this kind of smoothness constraint of the phase can also alleviate phase wrapping errors and noise to some degree. The size of the neighborhood area should be chosen according to the specific situation.

Let \( \phi(x_i) \) and \( \phi(\Omega_i) \) be the phase value of the \( i \)th pixel and the mean phase value in \( \Omega_i \), respectively. The smoothness constraint on the phase intensities is then defined as
\[
\| \phi(\mathbf{x}) - \hat{\phi}(\mathbf{x}) \|_2 = \left( \sum_{i=1}^{N^2} (\phi(x_i) - \phi(\Omega_i))^2 \right)^{1/2} \leq \sigma,
\]
with
\[
\phi(x_i) = \begin{cases} 
-\log \frac{x_i}{|x_i|}, & \text{if } |x_i| \geq T \in [-\pi, \pi). \\
0, & \text{otherwise}
\end{cases}
\]

Here the parameter \( \sigma \) is used to control the similarities between the \( i \)th pixel and its nearest neighbors, and \( T \) refers to a given threshold for separating the regions containing signal and noise only. Note that the reconstructed phase is arbitrary and carries no physical information at the location where the signal is very weak (i.e., the amplitude intensity is less than a given value \( T \)). Such artifacts will seriously affect the appearance of the restored results. We will show its effectiveness in detail in the experimental section.

In addition, many existing time-domain THz imaging systems, especially the one in Fig. 1, are based on a transmission-type spectroscopy. With no loss of generality, we assume that the object is piecewise homogeneous and has uniform thickness. Thus, we get one additional piece of prior knowledge about the original signals, which is that the smooth regions in the spatial distribution of amplitude should be the same as those in the spatial distribution of phase. Mathematically, let \( x_1 = A_1 \exp(i \phi_1) \) and \( x_2 = A_2 \exp(i \phi_2) \) be the complex intensities of two different pixels. If these two pixels are in the same homogeneous region, then \( A_1 = A_2 \) and \( \phi_1 = \phi_2 \). Accordingly, we can conclude that the first-order difference of the complex intensities in a homogeneous region is zero.

According to this assumption, if we define the total variation (TV) of a two-dimensional complex datum based on the \( \ell_0 \) norm as
\[
\sum_i \| \Delta_i^x \mathbf{x} \|_0 + \| \Delta_i^\phi \mathbf{x} \|_0,
\]
with \( \Delta_i^x \) and \( \Delta_i^\phi \) as linear operators corresponding to, respectively, horizontal and vertical first-order differences at pixel \( i \), then minimizing the TV will be a more appropriate choice for sparse complex image reconstruction in THz systems. Such a definition is not only good for signal sparsifying, but also emphasizes the correlation between the spatial distributions of the amplitude and phase. That is, for the solution that produces the minimum value of that \( \ell_0 \) norm TV, any two different signals in the same homogeneous region will not only have the same amplitude intensities, but also the same phase. The other optimization processes, such as minimizing the \( \ell_1 \) norm of the signal itself or the coefficients in a certain transform domain, however, cannot efficiently utilize such spatial correlation. Meanwhile, although some transform operation (e.g., wavelet transform) may sparsify the signal, the spatial correlation between the amplitude and phase will be destroyed in that transform domain.

For tractable programming, we replace \( \ell_0 \) norm with \( \ell_1 \) norm, i.e.,
\[
\| \mathbf{x} \|_{TV} = \sum_i (|\Delta_i^x \mathbf{x}|^2 + |\Delta_i^\phi \mathbf{x}|^2)^{1/2}.
\]

Such a form, which is defined as the complex-valued TV, also appears in [25], as holography is another common optical system that involves capturing complex signals [26, 27]. Then minimizing Eq. (7) will give us the same solution as the one by minimizing the \( \ell_0 \) norm TV. At this point, we would like to mention that in some sources [28], another form of the \( \ell_1 \) norm TV on complex data has been used as follows:
\[
\| \mathbf{x} \|_{TV} = \sum_i (|\Delta_i^x \mathbf{x}|^2 + |\Delta_i^\phi \mathbf{x}|^2)^{1/2}.
\]

However, the distinction between these two regularizers should be kept in mind, since, at least in our optimization problems, the definition in Eq. (7) leads to better results with much sharper edges, as illustrated in Section 3. Here and below, the term \( \| \mathbf{x} \|_{TV} \) refers to our first definition in Eq. (7).

Considering the case at a particular frequency \( f_i \), the sparse reconstruction algorithm for the CS THz imaging system can be interpreted as an optimization given by
\[
\begin{align*}
\text{minimize} & \quad \| \mathbf{x}(f_i) \|_{TV} \\
\text{subject to} & \quad \| \Psi \mathbf{x}(f_i) - \mathbf{b}(f_i) \|_2 \leq \epsilon, \\
& \quad \| \phi(\mathbf{x}(f_i)) - \hat{\phi}(\mathbf{x}(f_i)) \|_2 \leq \sigma,
\end{align*}
\]
or, equivalently, by the following criterion
\[
\hat{x}(f_i) = \arg \min_{x(f_i)} \frac{1}{2} \| \Psi x(f_i) - b(f_i) \|_2^2 + \lambda \| x(f_i) \|_{TV} + \mu \| \phi(x(f_i)) - \phi(\hat{x}(f_i)) \|_2^2.
\]
Such a solution of Eq. (10) can be found by convex programming methods, e.g., the nonlinear conjugate gradient method combined with backtracking line search.

C. Complex Multiscale Hyperspectral Image Estimation

Time-domain THz systems provide the capability to capture multiple spectral information in the far-infrared range. Recently, researchers have begun to pay more and more attention to THz hyperspectral imaging, such as security screening [14,15]. However, degradation (i.e., noise, blurring, etc.) of the observations acquired from hyperspectral imaging systems is always inevitable, and thus recovery of hyperspectral imagery from degraded observations becomes a vital task. A direct and naïve thought is to treat each spectral channel as an independent signal and then process them separately. The main shortcoming of such an approach is that it fails to account for the intercorrelated relationships present in the original hyperspectral signal. Researchers thus consider more sophisticated techniques to alleviate the degradation by taking not only the individual channels, but also the correlation across them, into account. One major approach is wavelet-based hyperspectral image estimation, because wavelet-based techniques can provide good spatial adaptivity by virtue of the well-localized property of wavelets in both time and frequency.

In [29,30], Atkinson et al. present a near-optimal hyperspectral image estimation scheme by addressing the different nature of spatial and spectral dimensions, which thus avoids the requirement of second-order signal statistics. In particular, based on some assumptions of the interchannel correlation, the hyperspectral signal can be approximately decorrelated in the spatial domain by using a 2-D discrete wavelet transform and the discrete Fourier transform (DFT) in the spectral domain. However, their work is essentially derived from the analysis of the hyperspectral Wiener filter, which is based on the assumption that the system is corrupted by additive white Gaussian noise. Besides, to choose a suitable wavelet basis to guarantee an optimal convergence rate, some a priori information about the underlying signals has to be provided, e.g., the degrees of smoothness.

Another class of hyperspectral image estimation approaches, which can be categorized as partition-based methods, automatically adapts to the signal function without any user input or a priori notion. This enjoys the advantages of wavelet-based techniques, such as fast convergence rates. The work by Willett and Nowak [31] shows a way for Poisson intensity estimation of a single-channel signal by using a penalized likelihood method on recursive dyadic partitions. Then Krishnamurthy and Willett [32] use such a partition-based multiscale Poisson intensity estimator in hyperspectral imaging reconstruction. While these methods are concerned with real-valued signal estimation from photon-limited observations corrupted by Poisson noise, in this paper we develop a partition-based multiscale estimation scheme for complex-valued hyperspectral signals gathered from time-domain THz imaging systems.

It is reasonable to think of a hyperspectral signal acquired from a time-domain THz imaging system as a three-dimensional dataset, where the first two dimensions correspond to the spatial locations and the third dimension indicates the index of each spectral band. Let $\mathbf{x}$ and $\mathbf{y}$ respectively denote the true complex hyperspectral signal and the distorted complex hyperspectral observation on a grid of size $N \times N \times K$, where $K$ represents the number of spectral bands. Our goal is to estimate $\mathbf{x}$ from $\mathbf{y}$ as accurately as possible. Note that in this subsection, the hyperspectral data $\mathbf{x}$ and $\mathbf{y}$ are of size $N \times N \times K$.

In accordance with [32], a key feature of the hyperspectral images is that the boundaries and singularities are located at the same spatial positions across all the spectral bands, no matter how bad the contrast or perceptibility is at some band. This is also true in our THz imaging scenario. In each spectral band, the underlying amplitude and phase intensities contain the same spatial boundaries in their respective domains. Therefore, in what follows we process the amplitude and phase separately. Specifically, we decompose the underlying signal $\mathbf{x}$ into the amplitude part $\mathbf{x}_A$ and phase part $\mathbf{x}_\phi$, and equally define $\mathbf{y}_A$ and $\mathbf{y}_\phi$ for the observation $\mathbf{y}$. Since the signals concerned here are basically spatially piecewise constant, to better adapt to the spatial distributions and preserve the discontinuities, we choose the multiscale approach by recursively performing recursive dyadic partitioning (RDP) in the amplitude and phase domains. The RDP process on an image produces a tree representation by recursively decomposing any part of an existing partition into dyadic squares, replacing a square by four similar squares of half the size [33]. Thus, the spatial RDPs can be represented in terms of quad-trees. Figure 2(a) gives a sample of an incomplete RDP and its tree. Since the partition defined by the RDP is not unique, we use penalized likelihood estimation to select the optimal partition $\tilde{\mathcal{P}}$ that provides the best fit to the observations from the space of possible partitions $\Sigma_{\mathcal{P}}$. Each of the terminal squares of this data-adaptive RDP $\tilde{\mathcal{P}}$ corresponds to a spatially homogeneous region. An example of such partitioning on a two-dimensional image is shown in Fig. 2(b).

In particular, at the beginning, complete spatial RDPs in amplitude and phase domains are obtained by recursively partitioning their respective spaces into cells with dyadic side length until reaching the pixel-level resolution. Then starting from this finest level, we search backward to find the fittest partition over every level of the quad-tree representation by merging adjoining

![Fig. 2. (Color online) (a) A Example of an incomplete RDP and its quad-tree. (b) Sample partition of a two-dimensional image.](image-url)
squares based on the similarities of the intensities in the neighboring cells. We let $\mathcal{P}_A$ and $\mathcal{P}_\phi$ denote the respective spatial RDP representations of amplitude and phase domains. Thus in each partition cell $c_\alpha \in \mathcal{P}_A$ and $c_\phi \in \mathcal{P}_\phi$, based on the respective likelihood of observing $y_A^{(c_\alpha)}$ and $y_\phi^{(c_\phi)}$ given the amplitude estimate $\hat{x}_A^{(c_\alpha)}$ and phase estimate $\hat{x}_\phi^{(c_\phi)}$, $p(y_A^{(c_\alpha)}|x_A^{(c_\alpha)})$ and $p(y_\phi^{(c_\phi)}|x_\phi^{(c_\phi)})$, we separately compute the following penalized likelihood (in a log-likelihood sense):

$$\hat{x}_A^{(c_\alpha)} = \arg \min_{x_A^{(c_\alpha)}} L_A^{(c_\alpha)} = \arg \min_{x_A^{(c_\alpha)}} \{- \log p(y_A^{(c_\alpha)}|x_A^{(c_\alpha)})\},$$

(11a)

$$\hat{x}_\phi^{(c_\phi)} = \arg \min_{x_\phi^{(c_\phi)}} L_\phi^{(c_\phi)} = \arg \min_{x_\phi^{(c_\phi)}} \{- \log p(y_\phi^{(c_\phi)}|x_\phi^{(c_\phi)})\}.$$  

(11b)

Apparently, the conditional probability functions $p(y_A^{(c_\alpha)}|x_A^{(c_\alpha)})$ and $p(y_\phi^{(c_\phi)}|x_\phi^{(c_\phi)})$ are directly related to the noise distributions in the particular THz system. While some work has been performed on modeling the distortions in the amplitude and phase domains [34], it is still an open problem in deciding an accurate noise estimation; in this paper, we choose the Gaussian distributions with mean zero and variance $\sigma_A^2$ and $\sigma_\phi^2$ for the likelihood estimation, i.e.,

$$p(y_A^{(c_\alpha)}|x_A^{(c_\alpha)}) = \prod_{(i,j) \in c_\alpha} \prod_{k=0}^{M-1} \frac{1}{\sqrt{2\pi\sigma_A^2}} \exp\left\{ -\frac{(y_A(i,j,k) - x_A(i,j,k))^2}{2\sigma_A^2} \right\},$$

(12a)

$$p(y_\phi^{(c_\phi)}|x_\phi^{(c_\phi)}) = \prod_{(i,j) \in c_\phi} \prod_{k=0}^{M-1} \frac{1}{\sqrt{2\pi\sigma_\phi^2}} \exp\left\{ -\frac{(y_\phi(i,j,k) - x_\phi(i,j,k))^2}{2\sigma_\phi^2} \right\}.$$  

(12b)

Certainly, $p(y_A^{(c_\alpha)}|x_A^{(c_\alpha)})$ and $p(y_\phi^{(c_\phi)}|x_\phi^{(c_\phi)})$ can be modified for other kinds of distribution in the cases where the noise can be accurately estimated. Anyhow, such an algorithm gives the users high design freedom. Finally, the optimal RDP representations in the amplitude and phase domains are respectively calculated by selecting the ones with the minimal total penalized likelihood, and then the respective optimal estimators for the amplitude and phase intensities are the ones fit best to the observations over every cell in that optimal RDP, i.e.,

$$\hat{\mathbf{x}}_A = \arg \min_{\mathbf{x}_A \in \mathcal{P}_A} \left\{ \sum_{c_\alpha \in \mathcal{P}_A} L_A^{(c_\alpha)} + \eta_A(\mathcal{P}_A) \right\},$$

(13a)

$$\hat{\mathbf{x}}_\phi = \left\{ \sum_{c_\phi \in \mathcal{P}_\phi} \hat{x}_\phi^{(c_\phi)} \right\},$$

(13b)

and

$$\hat{P}_\phi = \arg \min_{P_\phi} \left\{ \sum_{c_\phi \in \mathcal{P}_\phi} L_\phi^{(c_\phi)} + \eta_\phi(\mathcal{P}_\phi) \right\},$$

(14a)

$$\hat{\mathbf{x}}_\phi = \left\{ \sum_{c_\phi \in \mathcal{P}_\phi} \hat{x}_\phi^{(c_\phi)} \right\}. $$

(14b)

Note that in the above equations, the superscripts $c_\alpha$ and $c_\phi$ restrict the areas of interest, and the variables with these superscripts are zero outside the corresponding areas. The terms $\eta_A(\mathcal{P}_A)$ and $\eta_\phi(\mathcal{P}_\phi)$ are penalties for controlling the spatial smoothness in the amplitude and phase domains, respectively. The values of these two penalties are respectively proportional to the number of cells in the partitions $\mathcal{P}_A$ and $\mathcal{P}_\phi$. More technical details about the penalties can be found in [31]. Except for other hyperspectral image estimation methods, such as the ones mentioned in [29,30], the key distinction is the use of the correlated relationships in the hyperspectral images. In our method, such relationships are exploited by separately forcing the spatial RDPs of the amplitude and phase dimensions at each spectral band to be the same.

### D. Compressed Sensing hyperspectral image reconstruction

Incorporating the single-band sparse reconstruction approach presented in Subsection 2.B with the complex multiscale hyperspectral image estimation technique in subsection 2.C, we devise a hyperspectral image reconstruction algorithm for the CS THz imaging system. The algorithm flow chart is shown in Fig. 3. It consists of two alternating steps:

**Step 1:** Let $\hat{\mathbf{x}}^{(t)}$ be the input of the $t$th iteration. Obtain the solution $\hat{\mathbf{x}}^{(t)}$ by computing Eq. (10) over each observed spectral band.

**Step 2:** Considering $\hat{\mathbf{x}}^{(t)}$ as the observations in this estimation process, perform our complex multiscale hyperspectral estimation method, then obtain $\hat{\mathbf{x}}^{(t+1)}$. Specifically, deal with the amplitude and phase parts separately and get the respective optimal estimators $\hat{\mathbf{x}}_A^{(t+1)}$ and $\hat{\mathbf{x}}_\phi^{(t+1)}$ according to Eqs. (11)–(14). Combine the amplitude and phase and then obtain $\hat{\mathbf{x}}^{(t+1)}$.

These two steps are executed repeatedly, and the algorithm terminates when

$$\frac{\|\mathbf{x}^{(t+1)} - \hat{\mathbf{x}}^{(t)}\|_1}{\|\hat{\mathbf{x}}^{(t)}\|_1}$$

is small enough.

### 3. RESULTS AND DISCUSSION

#### A. Single-Band CS Reconstruction Results

We expand on the preliminary results first presented in [17] with further experiments and analyses. The experimental data are acquired with the help of Wai Lam (William) Chan and Daniel M. Mittleman of Rice University, who are the designers of the single-pixel THz imaging system [8]. The test object is a rectangular hole in an opaque screen filled with two transparent plastic plates of different thickness, as shown in Fig. 4. Figure 5 shows the CS reconstruction results by using different approaches...
only from the linear measurements acquired at a specific frequency (0.1 THz). In these experiments, 600 measurements are used to estimate the original complex signal of size $32 \times 32$.

We first perform the reconstruction by directly minimizing the $\ell_1$ norm of the underlying signal itself. The results are shown in Figs. 5(a) and 5(b). Figures 5(c) and 5(d) and Figs. 5(e) and 5(f) are the respective results by using two different wavelet functions (i.e., Daubechies-4 wavelet and Daubechies-8 wavelet) for sparsification. These results are obtained by using the Matlab code package of SPGL1 algorithm [20]. The SPGL1 algorithm in [19] is an often used method which is designed to solve the conventional sparse reconstruction problem expressed in Eq. (4). Moreover, the SPGL1 algorithm is one of a few approaches which can deal with the problems in the complex domain. Although the improvement is recognizable in comparison with the direct $\ell_1$ norm minimization method, the visual quality of these recoveries is still unsatisfactory. For example, one still cannot identify the edges of each plastic plate, and the artifacts distributed in the background of both amplitude and phase images are quite obvious.

Fig. 3. Hyperspectral image reconstruction algorithm flow chart.

Fig. 4. (Color online) Rectangular object mask [17].

Fig. 5. (Color online) CS reconstruction results on the single-band THz data at the frequency of 0.1 THz. Each image is of size $32 \times 32$. The left column shows the amplitude images, and the phase images are in the right column. (a), (b) Results obtained by directly minimizing the $\ell_1$ norm of the underlying signal itself. (c), (d) Results reconstructed by minimizing the $\ell_1$ norm of the Daubechies-4 wavelet coefficients. (e), (f) Results obtained by using Daubechies-8 wavelet transform for sparsification. (g), (h) Results obtained by applying our single-band CS reconstruction method with Eq. (7) as the complex TV norm. (i), (j) Results obtained by using our single-band approach with the TV norm in the form of Eq. (8).
Next, we show the recovered amplitude and phase images by performing our single-band CS reconstruction methods. Note that in Subsection 2.B, we introduce two different forms of the complex TV norm, i.e., Eqs. (7) and (8). Figures 5(g) and 5(h) correspond to using Eq. (7), and the results with the TV norm of Eq. (8) are shown in Figs. 5(i) and 5(j). From the results, we can easily see that our single-band algorithm with either of these two forms of TV norm outperforms the conventional method. Yet the distinction between these two different forms also needs to be dealt with. The recovery with the form of Eq. (7) yields images consisting of smoother regions with much sharper edges. Hence, in the following experiments, we choose it as our complex TV regularizer. In addition, in the phase smoothing regularizer, the smoothness degree is controlled by the size of the neighborhood area around each pixel. Since too large a window will cause oversmoothing in the estimation, accordingly we use a window of $3 \times 3$ pixels.

B. CS Hyperspectral Multiscale Reconstruction Results

In the experiments discussed below, we demonstrate the effectiveness of our proposed CS hyperspectral image reconstruction method on a set of practical hyperspectral THz measurements. The experimental platform is the same as in the previous experiment. Since each measurement obtained from the time-domain THz system is actually a whole pulse signal containing frequency information across the THz frequency range, we sample the measurements at 16 spectral bands uniformly distributed over the frequency range between 0.1 THz and 0.2 THz. We first conduct the experiments with 600 measurements at each spectral band. Therefore, our goal is to find the optimal estimation of the underlying hyperspectral data of size $32 \times 32 \times 16$ with only $600 \times 16$ linear measurements.

Figures 6(a) and 6(b) show the amplitude and phase images obtained by applying the conventional CS reconstruction method with the information only at 0.1 THz. Fig. 6. (Color online) Hyperspectral reconstruction results with the practical THz data of size $600 \times 16$. (a), (b) Amplitude and phase obtained by minimizing the $\ell_1$ norm of the Daubechies-8 wavelet coefficients only with the measurements at 0.1 THz. (c), (d) Amplitude and phase reconstructed by using our single-band reconstruction method only with the measurements at 0.1 THz. (e), (f) Amplitude and phase obtained by performing our proposed hyperspectral algorithm with data across all 16 spectral bands, displayed at 0.1 THz.

Next, we show the recovered amplitude and phase images by performing our single-band CS reconstruction methods. Note that in Subsection 2.B, we introduce two different forms of the complex TV norm, i.e., Eqs. (7) and (8). Figures 5(g) and 5(h) correspond to using Eq. (7), and the results with the TV norm of Eq. (8) are shown in Figs. 5(i) and 5(j). From the results, we can easily see that our single-band algorithm with either of these two forms of TV norm outperforms the conventional method. Yet the distinction between these two different forms also needs to be dealt with. The recovery with the form of Eq. (7) yields images consisting of smoother regions with much sharper edges. Hence, in the following experiments, we choose it as our complex TV regularizer. In addition, in the phase smoothing regularizer, the smoothness degree is controlled by the size of the neighborhood area around each pixel. Since too large a window will cause oversmoothing in the estimation, accordingly we use a window of $3 \times 3$ pixels.

Fig. 7. (Color online) Hyperspectral reconstruction results with the practical THz data of size $400 \times 16$. (a),(b) Amplitude and phase obtained by minimizing the $\ell_1$ norm of the Daubechies-8 wavelet coefficients only with the measurements at 0.1 THz. (c), (d) Amplitude and phase reconstructed by using our single-band reconstruction method only with the measurements at 0.1 THz. (e), (f) Amplitude and phase obtained by performing our proposed hyperspectral algorithm with data across all 16 spectral bands, displayed at 0.1 THz.
The amplitude and phase reconstructed by using our single-band reconstruction algorithm in Subsection 2.B at the same frequency are shown in Figs. 6(c) and 6(d). In agreement with our previous discussion, the latter performs much better than the conventional method. However, these methods fail to preserve some fine edges, and we still cannot easily estimate the object edges accurately.

Now let us compare our proposed hyperspectral algorithm, shown in Figs. 6(e) and 6(f), with two reconstruction approaches that are based only on the consideration of single-band information. Visually, the reconstruction quality is significantly improved in both the amplitude and phase dimensions, i.e., clearer contours with less noticeable artifacts and smoother in the homogeneous regions. More importantly, the amplitude and phase recoveries corresponding to our proposed hyperspectral method are more coincident with the correlation between the amplitude and phase (see Subsection 2.B) and also closer to the reality that the two plastic plates are homogeneous. In addition, since each pixel corresponds to a 1 mm area, we get the size of the outer rectangular by counting pixels (15 mm × 14 mm), which is equal to the actual measurement.

Furthermore, we also experiment with decreasing the number of the measurements to 400 × 16 (see Fig. 7). We can see that the results reconstructed from only the information at a single spectral band are still worse in quality, compared with our hyperspectral recovery. Our hyperspectral algorithm produces almost the same results, even though fewer measurements are used for reconstruction.

4. CONCLUSION

This paper studies methods for accurate image recovery in compressive THz imaging. First, we present a single-band CS reconstruction method. Unlike the conventional approach, which emphasizes the sparsity of only the amplitude, our single-band algorithm results in significant improvement by adding control on the phase and involving the correlation between the spatial distributions of amplitude and phase into the reconstruction process. Second, the hyperspectral nature of the THz pulse inspires us to devise a CS hyperspectral image reconstruction algorithm. In addition to the features of the single-band one, our hyperspectral method effectively employs correlations across the hyperspectral bands and shows good performance in preserving edges and alleviating artifacts in both amplitude and phase domains. Experimental results support these claims. Although the discussion in this paper is mainly based on the single-pixel THz imaging system in [8], many of contributions and discoveries can be readily utilized in other THz systems, such as time-domain THz spectroscopies.

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