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<th>Stabilization of systems with probabilistic interval input delays and its applications to networked control systems</th>
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Stabilization of Systems With Probabilistic Interval Input Delays and Its Applications to Networked Control Systems

Dong Yue, Engang Tian, Zidong Wang, and James Lam

Abstract—Motivated by the study of a class of networked control systems, this correspondence paper is concerned with the design problem of stabilization controllers for linear systems with stochastic input delays. Different from the common assumptions on time delays, it is assumed here that the probability distribution of the delay taking values in some intervals is known a priori. By making full use of the information concerning the probability distribution of the delays, criteria for the stochastic stability and stabilization controller design are derived. Traditionally, in the case that the variation range of the time delay is available, the maximum allowable bound of time delays can be calculated to ensure the stability of the time-delay system. It is shown, via numerical examples, that such a maximum allowable bound could be made larger in the case that the probability distribution of the time delay is known.

Index Terms—Delay, linear matrix inequality (LMI), Lyapunov functional, networked control system (NCS), stabilization.

I. INTRODUCTION

In many practical systems, such as networked and process control systems, the arrival of the measurement and control signals at the receiving point will often take some time, which leads to delayed input. For example, considering the transfer delays of sensor to controller, controller to actuator, and the effect of the data dropout, the linear networked control system (NCS) can be expressed as [31], [32]

\[ x(t) = A x(t) + B u(t - \tau(t)) \] (1)

where \( x(t) \in \mathbb{R}^n \) and \( u(t) \in \mathbb{R}^m \) denote the state and control vectors, respectively, \( A \) and \( B \) are constant matrices with appropriate dimensions, and \( \tau(t) \) denotes the time delay which is piecewise continuous and bounded.

In the past several decades, there has been much attention on the study of stability and stabilization of systems with state delay or control input delay [1]–[3], [5], [6], [8], [12], [13], [15]–[17], [20], [21], [26]. Among these references, delay-independent methods [6], [13], [16] or delay-dependent methods [5], [12], [17], [20], [21] were recommended by Associate Editor G. Calafiore. Grant Committee of Hong Kong under Grant HKU 7031/07P. This paper was supported in part by the Royal Society of the U.K., by the National Natural Science Foundation of China under Grants 60774060, 60474079, and 60834002, by the High Technology Project of Jiangsu Province BG2006042, and by the Research Grant Committee of Hong Kong under Grant HKU 7031/07P. This paper was recommended by Associate Editor G. Calafiore.

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Notation: \( \mathbb{R}^n \) denotes the \( n \)-dimensional Euclidean space, \( \mathbb{R}^{n \times m} \) is the set of real \( n \times m \) matrices, \( I \) is the identity matrix of appropriate dimensions, and \( \| \cdot \| \) stands for the Euclidean vector norm or spectral norm as appropriate. \( Z_{0+} \) denotes the set of nonnegative integers. The notation \( X > 0 \) (respectively, \( X \geq 0 \)) for \( X \in \mathbb{R}^{n \times n} \) means that the matrix \( X \) is a real symmetric positive definite (respectively, positive semidefinite). For a real matrix \( B \) and two real symmetric matrices \( A \) and \( C \) of appropriate dimensions, \( \begin{bmatrix} A & * \\ B & C \end{bmatrix} \) denotes a real symmetric matrix, where * denotes the entries implied by symmetry. \( E \{ \cdot \} \) denotes the mathematical expectation.

II. SYSTEM DESCRIPTION AND PRELIMINARIES

The following assumption is needed to state our main results. Assumption 1: There exist constants \( \bar{\tau}_1 \) and \( \bar{\tau}_2 \), where \( 0 \leq \bar{\tau}_1 \leq \bar{\tau}_2 \), such that either \( \tau(t) \in [0, \bar{\tau}_1] \) or \( \tau(t) \in (\bar{\tau}_1, \bar{\tau}_2] \). Furthermore, the probability distribution of \( \tau(t) \) taking values in \([0, \bar{\tau}_1]\) and \((\bar{\tau}_1, \bar{\tau}_2]\) is known a priori.
For the stochastic input delay $\tau(t)$, we define the following two random events:

\[
\mathcal{F}_1: \quad \tau(t) \in [0, \tilde{\tau}_1] \\
\mathcal{F}_2: \quad \tau(t) \in (\tilde{\tau}_1, \tilde{\tau}_2],
\]

Then, a random variable can be defined as follows:

\[
\delta(t) = \begin{cases} 
1, & \mathcal{F}_1 \text{ occurs} \\
0, & \mathcal{F}_2 \text{ occurs}.
\end{cases}
\]

Furthermore, define two functions $\tau_1: \mathbb{R}_+ \to [0, \tilde{\tau}_1]$ and $\tau_2: \mathbb{R}_+ \to [\tilde{\tau}_1, \tilde{\tau}_2]$ such that

\[
\tau_1(t) = \begin{cases} 
\tau(t), & \text{if } \delta(t) = 1 \\
\tilde{\tau}_1, & \text{if } \delta(t) = 0
\end{cases}, \\
\tau_2(t) = \begin{cases} 
\tilde{\tau}_1, & \text{if } \delta(t) = 1 \\
\tau(t), & \text{if } \delta(t) = 0.
\end{cases}
\]

Using $\delta(t)$ and $\tau_i(t) (i = 1, 2)$, (1) can be equivalently written as

\[
\dot{x}(t) = Ax(t) + \delta(t) Bu(t - \tau_1(t)) + (1 - \delta(t)) Bu(t - \tau_2(t)).
\]  

**Remark 1:** It can be shown that $\delta(t)$ is a Markovian process. For the following use, it is assumed that $\delta(t)$ follows an unknown but exponential distribution of switchings. Moreover, from the definitions of $\tau_1(t)$, $\tau_2(t)$, and $\delta(t)$, it is shown that the solution of (1) is also that of (2). Therefore, by checking the stability of (2), we can also deduce that of (1). If $\tau(t) \in [0, \tilde{\tau}_1]$ for any $t \in \mathbb{R}_+$, (2) becomes

\[
\dot{x}(t) = Ax(t) + Bu(t - \tau_1(t)),
\]

which is a commonly studied system with bounded time-varying input delay. If $\tau(t) \in (\tilde{\tau}_1, \tilde{\tau}_2]$ for any $t \in \mathbb{R}_+$, (2) becomes

\[
\dot{x}(t) = Ax(t) + Bu(t - \tau_2(t)),
\]

which is the system with interval time-varying input delay.

**Remark 2:** The introduction of $\delta(t)$ is motivated by the ideas in [22], [24], [25], and [29], where the signal transmission is subject to random communication delays. However, in this correspondence paper, the random variable $\delta(t)$ is used to account for the probabilistic nature of the time delay taking values in a certain interval. To our knowledge, this correspondence paper represents the first attempt to consider the probabilistic behavior of the time delays falling into a given interval.

**Assumption 2:** $\delta(t)$ is a Bernoulli distributed sequence with \(\text{Prob}[\delta(t) = 1] = E[\delta(t)] = \delta_0\), and \(\text{Prob}[\delta(t) = 0] = 1 - E[\delta(t)] = 1 - \delta_0\), where $0 \leq \delta_0 \leq 1$ is a constant.

**Remark 3:** From Assumption 2, it can be shown that $E[(\delta(t) - \delta_0)^2] = \delta_0(1 - \delta_0)$.

In this correspondence paper, the designed controller is of a linear type, i.e.,

\[
u(t) = Kx(t)
\]

where $K$ is a constant matrix to be determined later.

Under control law (3), the closed-loop system of (2) is

\[
\dot{x}(t) = Ax(t) + \delta(t) BK x(t - \tau_1(t)) + (1 - \delta(t)) BK x(t - \tau_2(t)).
\]  

Since (4) is a stochastic system, for the stability analysis of (4), the following definition is needed.

**Definition 1:** The system (4) is said to be exponentially stable in the mean-square sense if there exist constants $\alpha > 0$ and $\beta > 0$ such that

\[
E \left\{ \|x(t)\|^2 \right\} \leq e^{\alpha t} \sup_{-2\tau_2 \leq s \leq 0} E \left\{ \|\phi(s)\|^2 \right\}
\]

where $\phi(\cdot)$ is the initial function in (4) defined as $x(t) = \phi(t)$, $t \in [-\tau_2, 0]$.

Moreover, we also need the following lemmas. The proof of Lemma 1 can be obtained easily while Lemma 2 can be derived via Lemma 1 and Remark 4.

**Lemma 1:** Suppose $X_1$ and $X_2$ are two constant matrices of same dimensions. Then for a given $h \in [0, 1]$

\[
qX_1 + (1 - q)X_2 \leq 0
\]

holds for any $q \in [h, 1]$ if and only if

\[
X_1 < 0 \\
hX_1 + (1 - h)X_2 < 0.
\]

**Remark 4:** It can be seen that, when $q \neq h$ (i.e., $q \in (h, 1)$), (6) and (7) are just sufficient (but not necessary) conditions for guaranteeing (5) to hold for any $q \in (h, 1)$.

**Lemma 2:** Suppose $\tau_i(t) (i = 1, 2)$ are defined as previously and $\Phi_i, \Pi_i (i = 1, 2)$, $\Omega = \Omega^T$ are some constant matrices with appropriate dimensions, then

\[
\tau_1(t)\Phi_1 + (\tilde{\tau}_1 - \tau_1(t)) \Phi_2 + (\tau_2(t) - \tau_1(t)) \Pi_1 + (\tilde{\tau}_2 - \tau_2(t)) \Pi_2 + \Omega < 0
\]

if the following inequalities hold:

\[
\tau_1 \Phi_1 + (\tilde{\tau}_2 - \tau_2(t)) \Pi_1 + \Omega < 0 \\
\tau_1 \Phi_1 + (\tilde{\tau}_2 - \tau_2(t)) \Pi_1 + \Omega < 0 \\
\tau_1 \Phi_2 + (\tilde{\tau}_2 - \tau_2(t)) \Pi_2 + \Omega < 0 \\
\tau_1 \Phi_2 + (\tilde{\tau}_2 - \tau_2(t)) \Pi_2 + \Omega < 0.
\]

**III. STABILITY ANALYSIS**

We rewrite (4) as

\[
\dot{x}(t) = Ax(t) + \delta_0 BK x(t - \tau_1(t)) + (1 - \delta_0) \\
\times BK x(t - \tau_2(t)) + (\delta(t) - \delta_0) \\
\times [BK x(t - \tau_1(t)) - BK x(t - \tau_2(t))].
\]  

Next, we will derive some sufficient conditions for guaranteeing the exponential stability in the mean-square sense of (13).

Define

\[
y(t) = Ac(t)
\]

where $A$ and $\zeta^T(t)$ are shown at the bottom of the page. Then, (13) can be expressed as

\[
\dot{x}(t) = y(t) + (\delta(t) - \delta_0) B \zeta(t)
\]

where $B = [0 \quad BK \quad 0 \quad -BK \quad 0]$. 

\[
A = \begin{bmatrix} 
A & \delta_0 BK & (1 - \delta_0) BK & 0 & 0 \\
\end{bmatrix}
\]

\[
\zeta^T(t) = [x^T(t) \quad x^T(t - \tau_1(t)) \quad x^T(t - \tau_2(t)) \quad x^T(t - \tilde{\tau}_1) \quad x^T(t - \tilde{\tau}_2) \quad y^T(t)]
\]
Theorem 1: System (4) is exponentially stable in the mean-square sense if, for given constants \( \tilde{\tau}_1, \tilde{\tau}_2, \delta_0, \) and matrix \( K, \) there exist matrices \( P > 0, Q_i > 0, R_i > 0, Z_i > 0 \) for \( i = 1, 2, \) and \( T_j, \) such that the following LMIs hold:

\[
\begin{bmatrix}
\Xi_{11} & * & * \\
* & \Xi_{22} & * \\
* & * & \Xi_{33}
\end{bmatrix} < 0, \quad l = 1, 2, 3, 4
\]  
(16)

where

\[
\Xi_{21} = \begin{bmatrix} \tilde{\tau}_1 N^T \\ \tilde{\tau}_2 N^T \\ \hat{\tau}_1 M^T \\ \hat{\tau}_2 M^T \end{bmatrix}, \quad \Xi_{22} = \begin{bmatrix} \tilde{\tau}_1 N^T \\ \tilde{\tau}_2 N^T \\ \hat{\tau}_1 M^T \\ \hat{\tau}_2 M^T \end{bmatrix}, \quad \Xi_{31} = \begin{bmatrix} \tilde{\tau}_1 N^T \\ \tilde{\tau}_2 N^T \\ \hat{\tau}_1 M^T \\ \hat{\tau}_2 M^T \end{bmatrix}, \quad \Xi_{33} (\text{diag}) = \begin{bmatrix} -\tilde{\tau}_1 R_1 -\hat{\tau}_1 Z_1 \\ -(\tilde{\tau}_2 - \hat{\tau}_1) R_2 \\ -(\tilde{\tau}_2 - \hat{\tau}_1) Z_2 \end{bmatrix}
\]

and \( \Xi_{11}, \) \( \Xi_{31}, \) and \( \Xi_{33} \) are shown at the bottom of the page.

Proof: Construct a Lyapunov functional as

\[
V(x_t) = x^T(t) P x(t) + \int_{t-\tau_1}^{t} x^T(s) Q_1 x(s) \, ds
+ \int_{t-\tau_2}^{t} x^T(s) Q_2 x(s) \, ds
+ \int_{t-\tau_1}^{t} y^T(v) R_2 y(v) \, dv
\]

where \( P > 0, Q_i > 0, R_i > 0, \) for \( i = 1, 2. \) The infinitesimal operator \( \mathcal{L} \) of \( V(x_t) \) is defined as follows [18]:

\[
\mathcal{L} V(x_t) = \lim_{\Delta \to 0^+} \frac{1}{\Delta} \left\{ E \left( V(x_{t+\Delta}) | x_t \right) - V(x_t) \right\}
\]

Then, from (17) and (18) and by using the slack matrix method [9], we can obtain

\[
\mathcal{L} V(x_t) = 2 x^T(t) P y(t) - x^T(t-\tau_1) Q_1 x(t-\tau_1) + x^T(t) (Q_1 + Q_2) x(t) - x^T(t-\tau_2) x(t-\tau_2) + Q_2 x(t-\tau_2) + y^T(t) (\tilde{\tau}_1 R_1 + (\tilde{\tau}_2 - \hat{\tau}_1) R_2) y(t)
+ \delta_0 (1 - \delta_0) \xi^T(t) B^T (\tilde{\tau}_1 Z_1 + (\tilde{\tau}_2 - \hat{\tau}_1) Z_2) B \xi(t)
- \int_{t-\tau_1}^{t} y^T(s) R_1 y(s) \, ds
- \int_{t-\tau_2}^{t} y^T(s) R_2 y(s) \, ds
- \delta_0 (1 - \delta_0) \int_{t-\tau_1}^{t} \xi^T(s) B^T Z_1 B \xi(s) \, ds
- \delta_0 (1 - \delta_0) \int_{t-\tau_2}^{t} \xi^T(s) B^T Z_2 B \xi(s) \, ds
\]
where

\[ \gamma^T(t) = \begin{bmatrix} x^T(t) & x^T(t - \tau_1(t)) & x^T(t - \tau_2(t)) \end{bmatrix} \]

\[ S^T = \begin{bmatrix} S_1^T & S_2^T & S_3^T \end{bmatrix}. \]

Note that, from (15)

\[ -2\gamma^T(t)N \int_{t-\tau_1(t)}^t \dot{x}(s) \, ds \]

\[ \leq \tau_1(t) \gamma^T(t) N \left( R_1^{-1} + Z_1^{-1} \right) N^T \zeta(t) \]

\[ + \int_{t-\tau_1(t)}^t y^T(s) R_1 y(s) \, ds \]

\[ + \int_{t-\tau_1(t)}^t (\delta(s) - \delta_0)^2 \zeta^T(s) B^T Z_1 B \zeta(s) \, ds. \]

Similarly, we can show that

\[ -2\gamma^T(t)M \int_{t-\tau_1(t)}^t \dot{x}(s) \, ds \]

\[ \leq (\tau_2(t) - \tau_1(t)) \gamma^T(t) M \left( R_2^{-1} + Z_2^{-1} \right) M^T \zeta(t) \]

\[ + \int_{t-\tau_1(t)}^t y^T(s) R_1 y(s) \, ds \]

\[ + \int_{t-\tau_1(t)}^t (\delta(s) - \delta_0)^2 \zeta^T(s) B^T Z_1 B \zeta(s) \, ds. \]

\[ 2\gamma^T(t)W \int_{t-\tau_2(t)}^t \dot{x}(s) \, ds \]

\[ \leq (\tau_2 - \tau_2(t)) \gamma^T(t) W \left( R_2^{-1} + Z_2^{-1} \right) W^T \zeta(t) \]

\[ + \int_{t-\tau_2(t)}^t y^T(s) R_2 y(s) \, ds \]

\[ + \int_{t-\tau_2(t)}^t (\delta(s) - \delta_0)^2 \zeta^T(s) B^T Z_2 B \zeta(s) \, ds. \]

Combining (19)–(23) and taking the expectation on both sides of (19), we obtain

\[ E\{L(x_i)\} \]

\[ \leq E\left\{ \zeta^T(t) \left[ \Xi_{11} + \delta_0 (1 - \delta_0) B^T (\tau_1 Z_1 + (\tau_2 - \tau_1) Z_2) B \right. \right. \]

\[ \left. \left. + \tau_1(t) N \left( R_1^{-1} + Z_1^{-1} \right) N^T \right] \zeta(t) \right. \]

\[ + (\tau_1 - \tau_1(t)) M \left( R_1^{-1} + Z_1^{-1} \right) M^T \]

\[ + \tau_2(t) - \tau_1(t) T \left( R_2^{-1} + Z_2^{-1} \right) T^T \zeta(t) , \]

\[ + (\tau_2 - \tau_2(t)) W \left( R_2^{-1} + Z_2^{-1} \right) W^T \zeta(t) \right\}. \]

By Lemma 2, it can be seen that (16) with \( l = 1, 2, 3, 4 \) are sufficient conditions for guaranteeing that the right-hand side of (24) is less than zero. Moreover, it is easy to show that, under the conditions in (16), there exists a small enough constant \( \lambda > 0 \) such that the right-hand side of (24) is less than \( -\lambda I \). Therefore, from (24), we have

\[ E\{L(x_i)\} \leq -\lambda E\left\{ \zeta^T(t) \zeta(t) \right\}. \]

Then, using a method similar to that of [19], we can show that

\[ E\left\{ x^T(t) x(t) \right\} \leq \bar{\alpha} e^{-\lambda t} \sup_{-2\tau_2 \leq s \leq 0} E\left\{ ||\phi(s)||^2 \right\} \]

where \( \bar{\alpha} = (\alpha/\lambda_{\text{min}}(P)). \) \[ \blacksquare \]

Remark 5: In the special case, when \( \tau(t) \in [\bar{\tau}_1, \bar{\tau}_2] \) for any \( t \in \mathbb{R}_+, \) (4) can be expressed as

\[ \dot{x}(t) = Ax(t) + BK x(t - \tau_2(t)). \]

For (16) with \( l = 1, 2, \) letting \( \delta_0 = 0, \) deleting the rows and columns corresponding to \( \Xi_{31} \) and \( \Xi_{32} \) and to \( Z_1 \) \( (j = 1, 2), \) and deleting the second row and second column and replacing \( M_1 \) by \( N_j \) \( (i = 1, 3, 4, 5, 6), \) we can obtain a result, denoted as Theorem 1’, to check the exponential stability of (26). Due to this correspondence paper’s space limitation, the detailed statement of Theorem 1’ is omitted here.

IV. STABILIZATION CONTROLLER DESIGN

In this section, the design criteria for the feedback control gain \( K \) will be derived based on Theorem 1 and Remark 5.

Theorem 2: System (4) with \( K = Y X^{-1} \) is exponentially stable in the mean-square sense, if, for given constants \( \bar{\tau}_1, \bar{\tau}_2, \delta_0, \) and \( \rho_{\text{max}}(m = 2, 3, 4), \) there exist matrices \( P > 0, Q_1 > 0, R_1 > 0, \) \( Z_1 > 0 \) \( (i = 1, 2), \) \( X, Y, N_j, M_j, T_j, \) and \( W_j \) \( (j = 1, 2, 3, 4, 5, 6) \) of appropriate
dimensions such that the following matrix equalities hold:

\[
\begin{bmatrix}
\Gamma_{11} & * & * \\
\Gamma_{21} & \Gamma_{22} & * \\
\Gamma_{31} & 0 & \Gamma_{33}
\end{bmatrix} < 0, \quad l = 1, 2, 3, 4 \tag{27}
\]

where

\[
\Gamma_{21} = \begin{bmatrix}
\check{\tau}_1 \check{N}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{M}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{W}^T
\end{bmatrix}, \quad \Gamma_{22} = \begin{bmatrix}
\check{\tau}_1 \check{N}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{M}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{W}^T
\end{bmatrix}, \quad \Gamma_{31} = \begin{bmatrix}
\check{\tau}_1 \check{M}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{M}^T \\
(\check{\tau}_2 - \check{\tau}_1) \check{W}^T
\end{bmatrix}
\]

and \( \hat{\Xi}_{31} \) are shown at the bottom of the page.

**Proof:** First, note that (16) equals to

\[
\check{N}^T = \begin{bmatrix}
\check{N}^T_1 & \check{N}^T_2 & \check{N}^T_3 & \check{N}^T_4
\end{bmatrix}, \quad \check{M}^T = \begin{bmatrix}
\check{M}^T_1 & \check{M}^T_2 & \check{M}^T_3 & \check{M}^T_4 & \check{M}^T_5 & \check{M}^T_6
\end{bmatrix}, \quad \check{W}^T = \begin{bmatrix}
\check{W}^T_1 & \check{W}^T_2 & \check{W}^T_3 & \check{W}^T_4 & \check{W}^T_5 & \check{W}^T_6
\end{bmatrix}
\]

and \( \hat{\Xi}_{31} \) are shown at the bottom of the page.

Then, the solvability of (27) can be replaced by (27)', where (27)' is derived from (27) by changing \( \Gamma_{33} = \check{\Xi}_{33} \) to \( -G_i - G_i^{-1} \). To efficiently solve it, we may employ the cone-complementarity method [7].

By Schur complements, (29) is equivalent to

\[
\begin{bmatrix}
\check{Z}_i^{-1} X^{-T} \\
X^{-1} - G_i^{-1}
\end{bmatrix} > 0.
\]

Define \( \hat{Z}_i = \check{Z}_i^{-1}, \hat{G}_i = G_i^{-1}, \) and \( \hat{X} = X^{-1} \). Similar to [7], we provide the following linearization algorithm for the solvability of Theorem 3.
TABLE I
MAXIMUM ALLOWABLE $\bar{\tau}_2$

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<td>Jiang et al. [10]</td>
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</tbody>
</table>

Algorithm 1 (For Theorem 2)
Given constants $\rho_m$ ($m = 2, 3, 4$) and let $N$ denote the maximum number of iterations.

Step 1: Find a feasible solution $\{X, \hat{X}, \hat{Z}_1, \hat{Z}_2, \hat{G}_1, G_1, \hat{G}_2, G_2\}$ to LMIs (27)′ and

$$
\begin{bmatrix}
\dot{Z}_i & X^T \\
\hat{X} & G_i
\end{bmatrix} > 0 \quad \begin{bmatrix}
\dot{Z}_i & I \\
I & \hat{Z}_i
\end{bmatrix} \geq 0
$$

$$
\begin{bmatrix}
G_i & I \\
I & \hat{G}_i
\end{bmatrix} \geq 0, \quad (i = 1, 2)
$$

$$
\begin{bmatrix}
X & I \\
I & \hat{X}
\end{bmatrix} \geq 0.
$$

(30)

If no feasible solution, EXIT and try other constants $\rho_m$ ($m = 2, 3, 4$). Set $k = 0$.

Step 2: Solve the following minimization problem:

$$
\min \text{ tr} \left( X_k \hat{X} + \hat{X}_k X \right) + \sum_{i=1}^{2} \left( \dot{Z}_{ik} \hat{Z}_i + \dot{Z}_i \hat{Z}_{ik} + G_{ik} \hat{G}_i + \hat{G}_{ik} G_i \right)
$$

subject to LMIs (27)′ and (30).

Step 3: If $X^T \hat{Z}_2^{-1} X > G_i$ ($i = 1, 2$) are satisfied, a controller is given by $K = Y X^{-T}$. Otherwise, set $k = k + 1$. If $k < N$, go to Step 2; otherwise, EXIT (no feasible solution is found).

V. NUMERICAL EXAMPLES

In this section, two numerical examples will be employed to show the applications of our method. In Example 1, stability criteria in Section III will be used to solve the maximum allowable value of the time delay under different cases. The applications of criteria in this correspondence paper to NCSs will be shown in Example 2.

Example 1: Consider the system

$$
\dot{x}(t) = \begin{bmatrix}
-0.5 & -2 \\
1 & -1
\end{bmatrix} x(t) + \begin{bmatrix}
-0.5 & -1 \\
0 & 0.6
\end{bmatrix} x(t - \tau(t))
$$

(32)

we are now concerned with the following two cases of the time delay $\tau(t)$:

Case 1) $0 \leq \bar{\tau}_1 \leq \tau(t) \leq \bar{\tau}_2$.

Table I compares our results with existing ones.

Case 2) Probability distribution of $\tau(t)$ is known.

In this case, Table II shows the upper bounds of $\tau(t)$ for different values of $\delta_0$ and $\bar{\tau}_1$.

TABLE II
MAXIMUM ALLOWABLE $\bar{\tau}_2$

<table>
<thead>
<tr>
<th>$\delta_0$</th>
<th>0.7</th>
<th>0.8</th>
<th>0.9</th>
<th>0.99</th>
</tr>
</thead>
<tbody>
<tr>
<td>$\bar{\tau}_1$</td>
<td>0.1</td>
<td>2.4149</td>
<td>3.3373</td>
<td>5.6150</td>
</tr>
<tr>
<td>$\bar{\tau}_2$</td>
<td>0.4</td>
<td>1.9634</td>
<td>2.5469</td>
<td>3.9919</td>
</tr>
<tr>
<td>$\bar{\tau}_3$</td>
<td>0.7</td>
<td>1.4875</td>
<td>1.7586</td>
<td>2.3213</td>
</tr>
</tbody>
</table>

Example 2: Consider a feedback control system [33]

$$
\dot{x}(t) = \begin{bmatrix}
0 & 1 \\
0 & -0.1
\end{bmatrix} x(t) + \begin{bmatrix}
0 \\
0.1
\end{bmatrix} u(t)
$$

(33)

which is assumed that the sensor, controller, and actuator are connected by a common network medium. For a given linear controller $u(t) = [-3.75 -11.5] x(t)$, the asymptotic stability of closed-loop system was investigated [11], [12], [21], [31], [33] under the consideration of the effect of the network conditions, such as the transmission delay of the data packet. The upper bounds of the transmission delay for guaranteeing the asymptotic stability of the system are, respectively, $4.5 \times 10^{-4}$ [33], 0.0538 [21], 0.7805 [12], 0.8695 [31], and 0.9412 [11]. By using Theorem 1′, it can be computed that the allowable upper bound of the transmission delay is 1.0432.

As pointed out in [31], an NCS can also be transformed into a system with interval time-varying delay. When assuming that the lower bound of the delay is not zero, the allowable upper bounds of the delay are given in [11] for the system. It was computed that the upper bound of the delay is 0.9635 for the case of $\bar{\tau}_1 = 0.2$ [11] or 0.9916 for the case of $\bar{\tau}_1 = 0.3$ [32]. By using Theorem 1′, the obtained upper bounds of the delay are 1.0467 for the case of $\bar{\tau}_1 = 0.2$ and 1.0487 for the case of $\bar{\tau}_1 = 0.3$.

By using Theorem 2′ and considering $\rho_2 = 0.2$, $\rho_3 = 10$, we obtain that the upper bound of the time delay is 67 for the case of $\bar{\tau}_1 = 0$ and the corresponding feedback gain $K = [-0.0022 -0.0218]$. Similarly, by using the method developed in [31] and employing Theorem 1′ with $K = [-0.0022 -0.0218]$, it can be obtained that the allowable upper bound of the time delay is 608. Choosing $\rho_2 = 0.2$, $\rho_3 = 0.12$, and $\rho_3 = 20$ and using Algorithm 1, we can obtain the computation results for different values of $\bar{\tau}_1$ and $\delta_0 = 0.8$. Then, by using Theorem 1, we can finally compute the allowable upper bound of the time delay for the corresponding feedback gains (see Table III).

It can be found by Table III that, when the probability distribution of the time delay can be observed, using Theorem 2 can lead to a larger allowable upper bound of the delay than that using Theorem 2′ under the similar requirements for the system performance.
VI. CONCLUSION

In this correspondence paper, the stabilization problem has been investigated for a class of linear systems with stochastic input delay. Based on the information of probability distribution of the time delay, a new model of the system, which has probability-distribution-dependent parameter matrices, has been proposed. In terms of the new model, some stability and stabilization criteria for the system have been derived by solving a set of LMIs. To derive the LMI conditions for determining the stability and stabilizability of the system, the convexity of a matrix equation has been employed to give less conservative results. Our main results have been applied to a commonly employed NCS. Examples have shown that the proposed method can lead to less conservative results. Furthermore, if the probability distribution of the delay is known a priori, the allowable upper bound of the delay may be larger than those derived for the case when only the variation range of the delay is used.

REFERENCES


